TOSHIBA Bipolar Linear Integrated Circuit Silicon Monolithic

TA8215H,TA8215L

18W BTL x 2CH Audio Power Amplifier

The thermal resistance θ j-T of TA8215H and TA8215L package designed for low thermal resistance, has a high efficiency of heat radiation.

The temperature rise of chip can be reduced, and the influence from the degradation of the features due to the temperature rise at the high output can also be reduced.

This stereo audio power IC, designed for car audio use, has two built—in channels to reduce the characteristic difference between L and R channels.

In addition, the functions of stand-by and muting, and a variety of protection circuits are involved.

Features

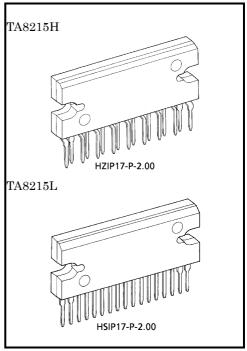
- High power
 - : P_{OUT} (1) = 18W (typ.) / channel (V_{CC} = 14.4V, f = 1kHz, THD = 10%, R_L = 4Ω)
 - : POUT(2) = 15W(typ.) / channel

$$(V_{CC} = 13.2V, f = 1kHz, THD = 10\%, R_L = 4\Omega)$$

- Low distortion ratio
 - : THD = 0.04% (typ.) $(V_{CC} = 13.2 V, \, f = 1 \text{kHz}, \, P_{OUT} = 1 W, \, R_L = 4 \Omega, \, G_V = 50 \text{dB})$
- Low noise
 - : $V_{NO} = 0.30 \text{mV}_{rms}$ (typ.) ($V_{CC} = 13.2 \text{V}$, $R_L = 4 \Omega$, $G_V 50 \text{dB}$, $R_g = 0 \Omega$, $BW = 20 \text{Hz} \sim 20 \text{kHz}$)
- Built-in stand-by function (with pin (4) set at low, power is turned off.)
 - $: I_{SB} = 1\mu A \text{ (typ.)}$
- Built-in muting function (with pin (1) set at low, power is turned off.)
 - $V_{(mute)} = 1V (typ.)$
- Built-in various protection circuits

Protection circuits: Thermal shut down, Over voltage, Out→V_{CC} short, Out→GND short and Out-Out short.

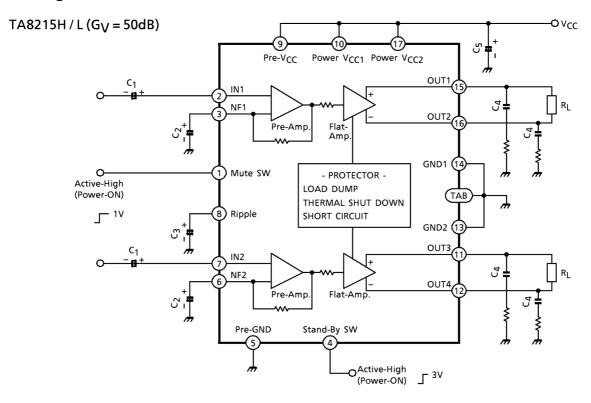
• Operating supply voltage : $V_{CC} = 9 \sim 18V$



Weight

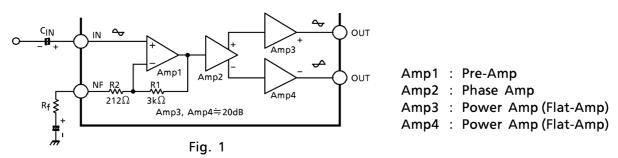
HZIP17-P-2.00 : 9.8g (typ.) HSIP17-P-2.00 : 9.8g (typ.)

Block Diagram



Cautions And Application Method (description is made only on the single channel.)

1. Voltage gain adjustment



This IC has the amplifier construction as shown in Fig.1. The Pre–Amp Amp1 is provided to the primary stage, and the input voltage is amplified by the Flat Amps, Amp3 and Amp4 of each channel through the phase Amp. (Amp2).

Since the input offset is prevented by Pre-Amp when V_{CC} is set to on, this circuit can remarkably reduce the popnoise.

The total closed loop gain GV of this IC can be obtained by expression below when the closed loop voltage gain of Amp1 is GV_1 .

$$G_{V1} = 20log \frac{R1 + (R_f + R2)}{R_f + R2} (dB) \dots (1)$$

The closed loop voltage gain of power Amp, Amp3 and Amp4 is fixed at $GV_3 \ = \ GV_4 = 20 dB$. Therefore, the total closed loop voltage gain GV is obtained through BTL connection by the expression

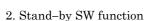
$$G_V = G_{V1} + G_{V3} + 6 \text{ (dB)} \quad \quad (2)$$

For example, when $R_f = 0\Omega$, GV is obtained by the expressions (1) and (2) as shown below.

$$GV = 24 + 20 + 6 = 50dB$$

The voltage gain is reduced when Rf is increased. (Fig.2)

With the voltage gain reduced, since (1) the oscillation stability is reduced, and (2) the pop noise changes when VCC is set to on, refer to the items 3 and 4.



By means of controlling pin (4) (stand–by terminal) to high and low, the power suply can be set to on and off. The threshold voltage of pin (4) is set at about 3V (typ.), and the power supply current is about $1\mu A$ (typ.) at the stand–by state.

(4) pin control voltage: V (SB)

Stand-by	Power	V (SB) (V)
On	Off	0~2
Off	On	3~V _{CC}

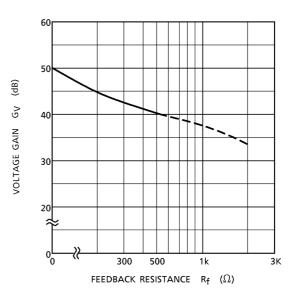


Fig. 2

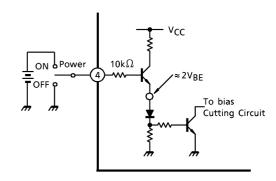


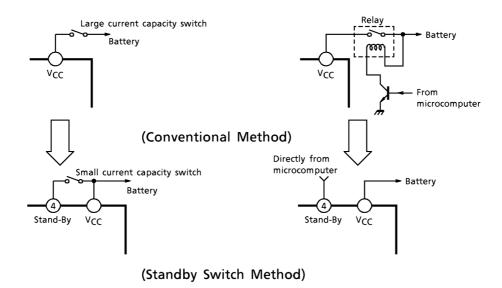
Fig. 3 With Pin⁽⁴⁾Set to High, Power is Turned ON.

~Advantage of stand-by SW~

- (1) Since VCC can directly be controlled to on, off by the microcomuter, the switching relay can be omitted.
- (2) Since the control current is microscopic, the switching relay of small current capacity is satisfactory for switching.

3

2002-10-30



3. Preventive measure against oscillation

For preventing the oscillation, it is advisable to use C₄, the condenser of polyester film having small characteristic fluctuation of the temperture and the frequency.

The resistance R to be series applied to C4 is effective for phase correction of high frequency, and improves the oscillation allowance.

Since the oscillation allowance is varied according to the causes described below, perform the temperature test to check the oscillation allowance.

- (1) Voltage gain to be used (Gy setting)
- (2) Capacity value of condenser
- (3) Kind of condenser
- (4) Layout of printed board

In case of its use with the voltage gain GV reduced or with the feedback amount increased, care must be taken because the phase–inversion is caused by the high frequency resulting in making the oscillation liable generated.

4. Input offset prevention circuit at V_{CC}→ON

Having Pre-Amp (Amp1) mounted on the primary stage, this IC contains the circuit for making the Amp1 input voltage and the NF terminal voltage equipotential.

Therefore, the offset voltage produced at the input stage is suppressed to prevent the pop noise at $V_{CC}\rightarrow ON$. The capacity values of the input and NF condense (C_1 and C_2) shall be set according to the gain to be used.

(Reference) (A) at
$$G_V = 50 dB$$
 ($R_f = 0 \Omega$) : $C_1 = 4.7 \mu F$, $C_2 = 47 \mu F$ (B) at $G_V = 40 dB$ ($R_f = 470 \Omega$) : $C_1 = 3.3 \mu F$, $C_2 = 33 \mu F$

5. Muting function

Through setting pin (1) (mute terminal) at about 1V or less, muting becomes possible.

The interval circuit of IC is shown in Fig.4.

When pin (1) is set to low, Q1 and Q2 are turned on, the charge of the ripple condenser is discharged and the bias is cut. The mute amount of 60dB or over can be obtained.

Since this muting function rapidly discharges the charge of the ripple filter capacitor of pin (8), the pop noise is generated by the DC fluctuation of the bias section.

Therefore, this muting function is not appropriate to the audio muting but is effective in muting at $V_{CC} \rightarrow on$.

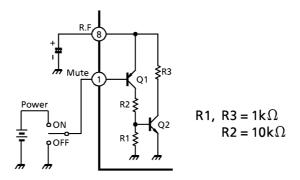


Fig. 4 Mute Circuit

6. External part list and description

Recommended			Influence			
Symbol Value	Feature	Smaller Than Recommended Value	Larger Than Recommended Value	Remarks		
C ₁	4.7µF	DC blocking	Related to pop noise at V _{CC} →on.		Related to gain. Refer to item 4.	
C ₂ 47µF	Feedback condenser	Related to pop noise at V _{CC} →on.				
		Determination of low cut-off frequency.				
		$C_2 = \frac{1}{2\pi f_L \cdot R_f}$		_		
C ₃	220μF	Ripple reduction	Time constant is small at V _{CC} →on or off.	Time constant is large at V _{CC} →on or off.	_	
C ₄	0.12µF	Oscillation prevention	Made liable to oscillate.	Oscillation allowance improved.	Refer to item 3.	
C ₅	1000μF	Ripple filter	For filtering power supply hum and ripple. Large at using AC rectified power supply. Small at using DC power supply.		_	

Maximum Ratings (Ta = 25°C)

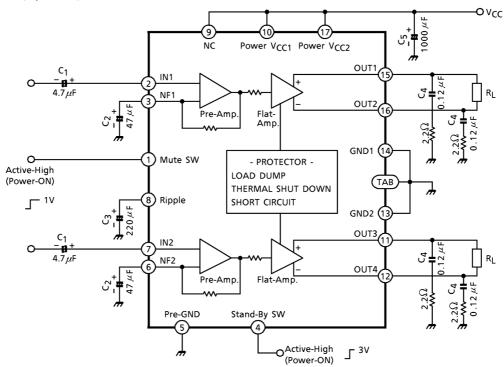
Characteristic	Symbol	Rating	Unit
Peak supply voltage (0.2s)	V _{CC} surge	50	V
DC supply voltage	V _{CC} DC	25	V
Operating supply voltage	V _{CC opr}	18	V
Output current (peak)	I _{O (peak)}	9	Α
Power dissipation	P _D	50	W
Operating temperature	T _{opr}	-30~85	°C
Storage temperature	T _{stg}	-55~150	°C

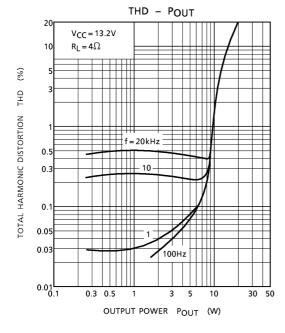
Electrical Characteristics (Unless otherwise specified, V_{CC} = 13.2V, R_L = 4 Ω , f = 1kHz, Ta = 25°C)

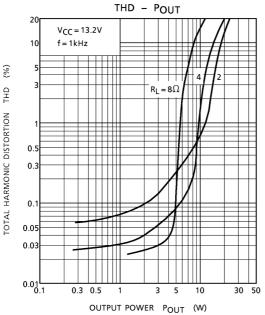
Characteristic	Symbol	Test Cir- cuit	Test Condition	Min.	Тур.	Max.	Unit	
Quiescent supply current	Iccq	_	V _{IN} = 0	_	120	250	mA	
Output power	P _{OUT} (1)	_	V _{CC} = 14.4V, THD = 10%	_	18	_	W	
	P _{OUT} (2)	_	THD = 10%	11	15	_		
Total harmonic distortion	THD	_	P _{OUT} = 1W	_	0.04	0.4	%	
Voltage gain	G _V	_	_	48	50	52	dB	
Output noise voltage	V _{NO}	_	$R_g = 0\Omega$, BW = 20Hz~20kHz	_	0.30	0.70	mV _{rms}	
Ripple rejection ratio	R.R.	_	fripple = 100 Hz, Rg = 600Ω	40	54	_	dB	
Input resistance	R _{IN}	-	_	-	30	_	kΩ	
Output offset voltage	Voffset	_	V _{IN} = 0	-0.3	0	0.3	V	
Current at stand-by state	I _{SB}	_	_	_	1	10	μA	
Cross talk	C.T.	_	$R_g = 600\Omega,$ $V_{OUT} = 0.775V_{rms} (0dBm)$	_	60	_	dB	
Pin (4) control voltage	V _(SB)	_	Stand−by → off (power→on)	2.5	_	V _{CC}	V	
Pin (1) control voltage	V _(MUTE)		Mute→on (power→off)		1.0	2.0	V	

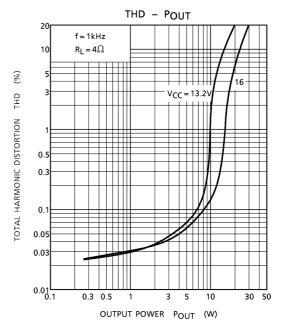
Test circuit

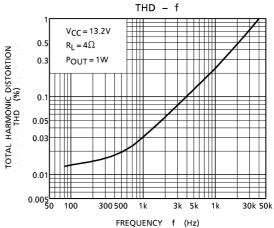
TA8215H / L ($G_V = 50dB$)

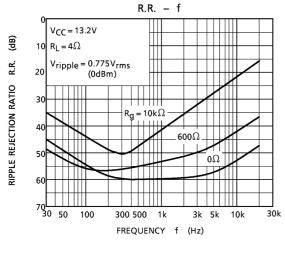


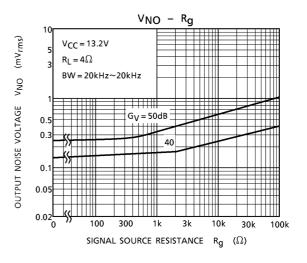


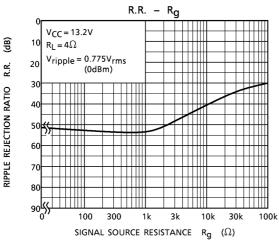


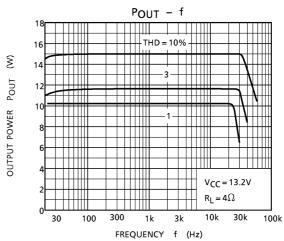


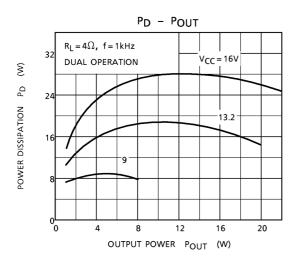


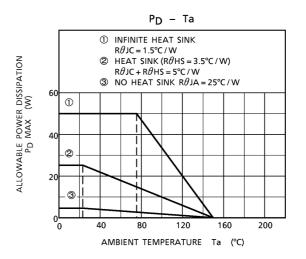












Unit: mm

Package Dimensions

HZIP17-P-2.00

5.0±0.3 30.0±0.2 14.0±0.2 $3.9_{-0.2}^{+0.2}$ 17.0±0.3 10.0±0.3 5.0±0.3 2.575TYP $0.4^{+0.1}_{-0.05}$ 0.55±0.1 ⊕ Ø 0.4 ₩ 2.0TYP 1.2±0.1 2.0 2.0 36.5MAX 36.0±0.2

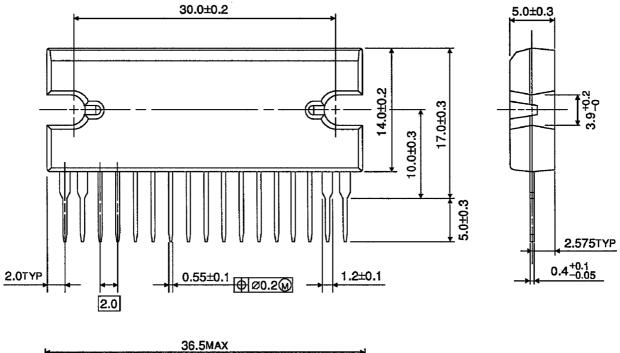
17

Weight: 9.8g (typ.)

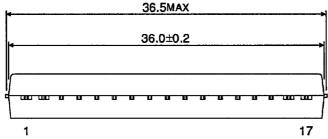
Package Dimensions

TOSHIBA

HSIP17-P-2.00 Unit: mm



11



Weight: 9.8g (typ.)

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